

- 1 -

NARROW-DIRECTIVITY ELECTROMAGNETIC-FIELD ANTENNA PROBE,  
AND ELECTROMAGNETIC-FIELD MEASUREMENT APPARATUS,  
ELECTRIC-CURRENT DISTRIBUTION SEARCH-FOR  
APPARATUS OR ELECTRICAL-WIRING DIAGNOSIS  
5 APPARATUS USING THIS ANTENNA PROBE

#### BACKGROUND OF THE INVENTION

The present invention relates to a probe and  
apparatuses using this probe for measuring proximate  
electromagnetic fields in proximity to high-frequency  
10 operating electronic appliances, information processing  
terminals, communications appliances, semiconductors,  
circuit boards, and the like, or for irradiating these  
targets with an electromagnetic field.

Conventionally, a small monopole antenna or a  
15 small loop antenna has been used as the probe, thereby  
performing the measurement of the electromagnetic  
fields or the irradiation with the electromagnetic  
field. As a result, it has been a limit to acquire a  
position resolution that is almost identical to a  
20 measurement height or an irradiation height, i.e., a  
spacing between a target to be measured and the probe.

In JP-A-2001-255347, the conventional  
proximate electromagnetic-field measuring probe has  
been disclosed as follows: In order to shield  
25 extraneous noises, it is selected as an object to

provide a proximate electromagnetic-field measuring antenna having unidirectionality. Moreover, in order to accomplish this object, the antenna is designed to be a one whose directionality is formed into the

5 unidirectionality by equipping the antenna with a metallic horn or a dielectric. This design makes the directionality unidirectional in the aperture direction of the metallic horn. Also, the existence of this metallic horn shields the extraneous noises.

10 Accordingly, it becomes possible to measure only a desired electromagnetic field.

#### SUMMARY OF THE INVENTION

When using the conventional small monopole antenna or the conventional small loop antenna as the

15 probe, the half-width of the probe is equal to substantially  $90^\circ$  and, considering the parallel component with a target to be measured, the half-width becomes equal to substantially  $45^\circ$ . Accordingly, the measurement-position resolution becomes almost

20 identical to the measurement height, since the probe height and the half-width become regions that are almost identical to each other. On account of this, there has existed the following problem: Unless the probe height is lowered by bringing the probe extremely

25 closer to the to-be-measured target, it is impossible to wish the implementation of enhancing the measurement-position resolution up to a higher

resolution.

Also, in the antenna disclosed in JP-A-2001-255347, the electric-current direction flowing in the main device and the electric-current direction flowing  
5 in the shield unit are in a mutually orthogonal relationship. As a result, the antenna exhibits an effect of shielding the main device from an electric field arriving thereat from a side above the shield-unit's lower surface. The antenna, however, has  
10 canceled out radiation electric-field components radiated toward a side below the shield-unit's lower surface, thereby finding it impossible to narrow the directionality. Consequently, there has existed the following problem: It is impossible to narrow, down to  
15 smaller than, the directionality of a radiation electric field radiated from the main device to the probe's lower portion.

In order to solve the above-described problems, it is required to narrow the directionality  
20 of the probe using the small monopole antenna or the small loop antenna. This makes it possible to acquire the position resolution that is higher than the probe height. For implementing this requirement, it is selected as an object to narrow the directionality of  
25 the small monopole antenna or that of the small loop antenna.

Other objects, features and advantages of the invention will become apparent from the following

description of the embodiments of the invention taken in conjunction with the accompanying drawings.

#### BRIEF DESCRIPTION OF THE DRAWINGS

FIG. 1 is a drawing for illustrating a  
5 narrow-directivity probe embodiment 1;

FIG. 2 is a drawing for illustrating a conventional-type probe;

FIG. 3 is a drawing for illustrating a narrow-directivity probe embodiment 2;

10 FIG. 4 is a drawing for illustrating a narrow-directivity probe device arrangement 1;

FIG. 5 is a drawing for illustrating a narrow-directivity probe device arrangement 2;

15 FIG. 6 is a drawing for illustrating an electric-field-type narrow-directivity probe embodiment 1;

FIG. 7 is a drawing for illustrating an in-plane electromagnetic-field intensity distribution generated by the conventional-type probe;

20 FIG. 8 is a drawing for illustrating an in-plane electromagnetic-field intensity distribution generated by the narrow-directivity probe embodiment 1;

FIG. 9 is a drawing for illustrating an in-plane electromagnetic-field intensity distribution  
25 generated by the narrow-directivity probe embodiment 2;

FIG. 10 is a drawing for illustrating a narrow-directivity probe embodiment 3;

FIG. 11 is a drawing for illustrating a narrow-directivity probe embodiment 4;

FIG. 12 is a drawing for illustrating an electromagnetic-field distribution

5 measurement/electric-current distribution search apparatus;

FIG. 13 is a drawing for illustrating an electromagnetic-field irradiation-type inspection apparatus;

10 FIG. 14 is a drawing for illustrating a pin-point electromagnetic-field generation mechanism embodiment 1 by a narrow-directivity probe array; and

FIG. 15 is a drawing for illustrating a pin-point electromagnetic-field generation mechanism  
15 embodiment 2 by the narrow-directivity probe array.

#### DETAILED DESCRIPTION OF THE EMBODIMENTS

Hereinafter, referring to the drawings, the explanation will be given below concerning embodiments of the present invention.

20 A conventional-type probe 200 illustrated in FIG. 2 extracts a signal line 103, and forms a main probe 101, and is a loop-shaped probe connected to grounds 104. In this shape, as a characteristic of the 1-wound small loop antenna, if, as illustrated in FIG.  
25 7, the probe exists on the yz plane, an in-xy-plane electromagnetic-field intensity distribution 701 generated thereby exhibits a comparatively rolling

distribution. On account of this, a region in which this in-xy-plane electromagnetic-field intensity distribution 701 becomes equal to a half-value with respect to the peak thereof, i.e., a position  
5 resolution at the time of a measurement, is of substantially the same order as the height of the probe. In view of this situation, in order to narrow this region and enhance the position resolution, a probe having a structure as illustrated in FIG. 1 is  
10 proposed.

A probe embodiment 1 illustrated in FIG. 1 extracts the signal line 103, and forms the main probe 101, and, as directionality-adjusting devices 102, simultaneously forms loop antennas that are inversely  
15 wound with respect to the main probe. Moreover, the respective lines are connected to the grounds 104. At this time, an electric-current path 105 of the main probe 101 and electric-current paths 106 of the directionality-adjusting devices 102 become opposite in  
20 their directions. As a result, even if identical-phase electric currents are fed thereto, electromagnetic fields generated thereby become opposite in their phases. On account of this, the electromagnetic fields generated by the directionality-adjusting devices 102  
25 operate such that these electromagnetic fields cancel out the electromagnetic field generated by the main probe 101. If, for example, the summation of the areas of the directionality-adjusting devices 102 is smaller

as compared with the area of the main probe 101, as illustrated in FIG. 8, an in-plane electromagnetic-field intensity distribution 801 finally generated becomes narrower as compared with the electromagnetic-field intensity distribution 701 generated by the conventional-type probe. This indicates that a narrow-directivity probe has been implemented.

Furthermore, in a probe embodiment 2 illustrated in FIG. 3, the directionality-adjusting devices 102 are located in a symmetric manner, i.e., located in the axis direction of the main probe 101 and in the direction perpendicular thereto. As illustrated in FIG. 9, this location implements, from the electromagnetic-field intensity distribution generated by the main probe 101, an electromagnetic-field intensity distribution 901 that is narrower than the electromagnetic-field intensity distribution 801 shown in the probe embodiment 1. This indicates that the probe embodiment 2 has become a narrow-directivity probe.

In this way, when the directionality-adjusting devices 102 are located around the main probe 101, the resultant electromagnetic-field intensity distribution can be focused in comparison with the case of the main probe 101 alone. This means that a narrow-directivity probe has been implemented. FIG. 4 illustrates a conceptual diagram thereof. Here, assuming that the electric-current path 105 of the main

probe 101 and the electric-current paths 106 of the directionality-adjusting devices 102 are identical in their directions, the fed electric-current phase difference between the main probe 101 and the

5 directionality-adjusting devices 102 located around the main probe 101 is shifted by  $\pi$  [rad]. This allows the directionality-adjusting devices 102 to cancel out the electromagnetic field generated by the main probe 101, thereby making it possible to narrow the

10 directionality. Meanwhile, as the embodiment illustrated in FIG. 1 or FIG. 3, even if the fed electric-currents are identical in their phases, basically the same result can be acquired as long as the electric-current path 105 of the main probe 101 and

15 the electric-current paths 106 of the directionality-adjusting devices 102 are opposite in their directions. Also, when the electric-current path 105 of the main probe 101 and the electric-current paths 106 of the directionality-adjusting devices 102 are identical in

20 their directions, the phase difference therebetween need not be completely equal to  $\pi$  [rad], but is allowable as long as the phase difference falls in the range of  $\pi \pm \pi/2$  [rad]. From this condition, when the electric-current path 105 of the main probe 101 and the

25 electric-current paths 106 of the directionality-adjusting devices 102 are opposite in their directions, the phase difference between the fed electric-currents is allowable up to a phase difference of  $0 \pm \pi/2$  [rad].



An object of these narrow-directivity probes is to focus the electromagnetic-field intensity distribution in the plane. These narrow-directivity probes, however, are of the symmetric shapes. This condition generates basically the same electromagnetic-field intensity distributions in a direction opposite to the observation plane as well, i.e., in the upward direction in the probe's configuration drawing illustrated in FIG. 4. In contrast thereto, as illustrated in FIG. 5, an adjustment device 501 whose directionality is antisymmetric is located above the main probe 101. This condition allows the probe's directionality to be focused in the observation-plane direction.

In the explanation given so far, the explanation has been given by selecting, as the central subject, the probes for focusing the magnetic-field intensity distribution and by referring to the drawings all of which use the loop antennas. As illustrated in FIG. 6, however, the use of monopole antennas also allows a narrow-directivity probe to be similarly implemented for an electric-field intensity distribution: Namely, directionality-adjusting devices 602 are located such that the devices 602 cancel out the electric-field intensity distribution generated by a main probe 601. In this case as well, as illustrated in FIG. 6, if the electric-current path directions are opposite ones, the phase difference between fed

electric-currents is allowable up to the phase difference of  $0 \pm \pi/2$  [rad]. Also, if the directions of the directionality-adjusting devices 602 are inverted, the phase difference between the fed  
5 electric-currents is allowable up to the phase difference of  $\pi \pm \pi/2$  [rad].

Next, referring to FIG. 10 and FIG. 11, the explanation will be given below concerning different embodiments of the configuration mode of the narrow-  
10 directivity probe. This configuration is as follows: As illustrated in FIG. 10, in a loop-shaped probe that extracts the signal line 103, and forms the main probe 101, and is connected to the grounds 104, there is provided a method of using conductor plates as the  
15 wiring of the grounds 104 to form the conductor plates into directionality-adjusting conductor plates 1001. Here, it has been known that, if an infinite conductor flat-plate exists for an electric current, a mirror image is configured at a position that is symmetrical  
20 to the plane. In this embodiment, the size of these conductor plates is made finite, thereby forming mirror images in an incomplete manner so as to substitute the directionality-adjusting conductor plates 1001 for the directionality-adjusting devices 102 illustrated in  
25 FIG. 1. Here, the condition that the conductor plates 1001 are required to satisfy is as follows: The directionality-adjusting conductor plates 1001 are larger than the main probe 101 so that, if the main

probe 101 is projected in the axis direction thereof,  
the entire main probe 101 can be projected on the  
plates 1001. This is because the plates 1001, although  
in the incomplete manner, are required to configure the  
5 mirror images. Here, in the narrow-directivity probe  
embodiment 3 (1000) illustrated in FIG. 10, as is the  
case with the narrow-directivity probe embodiment 1  
(100) illustrated in FIG. 1, an in-plane  
electromagnetic-field intensity distribution generated  
10 thereby becomes basically the same as the in-plane  
electromagnetic-field intensity distribution 801  
illustrated in FIG. 8. In view of this situation, as  
illustrated in FIG. 11, these directionality-adjusting  
conductor plates 1001 are connected to each other,  
15 thereby configuring a rectangular-parallelepiped shape.  
This configuration allows the directionality-adjusting  
conductor plates 1001 to be substituted for the  
directionality-adjusting devices 102 illustrated in  
FIG. 3. Accordingly, in this narrow-directivity probe  
20 embodiment 4 (1100), as is the case with the narrow-  
directivity probe embodiment 2 (300) illustrated in  
FIG. 3, an in-plane electromagnetic-field intensity  
distribution generated thereby becomes basically the  
same as the in-plane electromagnetic-field intensity  
25 distribution 901 illustrated in FIG. 9. In this way,  
the utilization of the mirror-image effect makes it  
possible to cause the directionality-adjusting  
conductor plates 1001 to play a role of the

directionality-adjusting devices 102. As the shape of the directionality-adjusting conductor plates 1001 in this case, in addition to the parallel flat-plates shape in FIG. 10 and the rectangular-parallelepiped shape in FIG. 11, various configurations such as a cylindrical shape are available. The condition for permitting the conductor plates 1001 to be substituted for the directionality-adjusting devices 102 is that the conductor plates 1001 have enough areas for permitting the main probe 101 to be projected in at least two directions.

The methods explained so far make it possible to configure the narrow-directivity probes. However, in the case of a configuration of having the maximum sensitivity in the front-side direction of the main probe 101, the following conditions are necessary: The directionality-adjusting devices 102 or the directionality-adjusting conductor plates 1001 are located at positions that are symmetrical to each other with respect to the main probe 101. Moreover, in order that each of the located directionality-adjusting device 102 or directionality-adjusting conductor plate 1001 will generate an electromagnetic field of the same magnitude, electric currents of the same magnitude are caused to flow in the devices 102 or the conductor plates 1001 which are in the above-described position-symmetry relationship, or the products of these electric currents are equal to each other, or the like.

In this case, however, the maximum sensitivity always exists on a line in the maximum-sensitivity direction. This condition results in the following problems: If an obstructing object exists  
5 halfway on the way to a target to be measured, it is impossible to perform the irradiation with an electromagnetic field in this direction here. Otherwise, if electromagnetic-wave sources exist, it is impossible to observe a desired electromagnetic-wave  
10 source. In view of this situation, as illustrated in FIG. 14, a plurality of narrow-directivity probes are prepared, and are located such that their maximum-sensitivity directions intersect with each other at a certain single point. As the result of this location,  
15 layer-basis in-plane electromagnetic-field intensity distributions 1401 in correspondence with distances from the plurality of probes have the maximum sensitivities at the point of the intersection. This allows the implementation of the electromagnetic-field  
20 irradiation at a pin point, or that of the observation of an electromagnetic-wave source.

Here, in FIG. 14, each of the narrow-directivity probes has been oriented to the desired position at which each of the maximum sensitivities is  
25 wished to be acquired. Tilting the maximum-sensitivity directions of the narrow-directivity probes, however, makes it possible to implement a configuration where the maximum-sensitivity directions are oriented to a

desired single point although, seemingly, the narrow-directivity probes are arranged within a certain plane. This tilting is implemented by reducing the sizes or the electric currents of the directionality-adjusting devices 102 or directionality-adjusting conductor plates 1001 located such that each maximum-sensitivity direction of each narrow-directivity probe is oriented to the desired direction. Otherwise, this tilting is implemented by reducing both the sizes and the electric currents. Furthermore, even if the sizes or the electric currents of the directionality-adjusting devices 102 or directionality-adjusting conductor plates 1001 are equal to each other, shifting the phases of the fed electric-currents allows the maximum-sensitivity directions to be tiled in the desired direction.

This makes it possible to configure a probe system having its maximum sensitivity at a 3-dimensionally desired position that is not limited within a plane.

The narrow-directivity probe 1203 explained so far is applicable to an apparatus 1200 illustrated in FIG. 12. The apparatus 1200 measures the electromagnetic-field distribution of an electronic appliance or the like, or searches for the electric-current distribution thereof from its result. This apparatus 1200 is configured by mounting the narrow-directivity probe 1203 on a 2/3/4-dimensional stage.

The apparatus 1200 scans the proximity to a to-be-measured target 1202, then measuring the distribution of the proximate magnetic field and/or electric field. Here, the apparatus 1200 has an antenna control circuit 5 1205 that includes a switch used as follows: In order to perform the rough measurement at first, and then in order to perform the detailed measurement of a location where the electric-field or magnetic-field component is intense or the like, the switch is used at first for 10 cutting off the directionality-adjusting devices 102 of the narrow-directivity probe 1203 to transform the narrow-directivity probe into a common probe, and, at the time of the detailed measurement, the switch is used for transforming the common probe back to the 15 narrow-directivity probe. This antenna control circuit 1205 is controlled using a computer 1211 or the like. Also, a signal induced by the probe 1203, depending on its intensity, is caused to pass through a high-frequency amplifier 1206, then being measured by a 20 measurement device 1210. At this time, in order to measure the phase component of this electromagnetic field as well, the following measurement steps are executed: The fundamental clock of the to-be-measured target 1202 is detected using a probe 1207 for 25 detecting the fundamental clock of the to-be-measured target 1202. Next, this signal is caused to pass through a frequency-dividing circuit 1208 and a frequency-multiplying circuit 1209 controlled using the

computer 1211 or the like, thereby being converted into a desired frequency component. Moreover, the synchronous detection with this desired frequency component is performed using the detected fundamental  
5 clock, thereby making it possible to measure the above-described phase component.

Also, the narrow-directivity probe 1203 is applicable to a test apparatus 1300 illustrated in FIG. 13. The test apparatus 1300, which is a test apparatus  
10 of an electronic appliance or the like, irradiates the electronic appliance or the like with an electromagnetic field. The apparatus 1300 is configured by mounting the narrow-directivity probe 1203 on the 2/3/4-dimensional stage. The apparatus  
15 1300 scans the proximity to the to-be-tested target 1202, then irradiating the to-be-tested target 1202 with an electromagnetic field from the proximity thereto. The narrow-directivity probe 1203 receives electric-power supply from a signal oscillator 1301,  
20 then irradiating a desired position on the to-be-tested target 1202 with the electromagnetic field. Here, as is the case with the apparatus 1200 for measuring the above-described electromagnetic-field distribution or searching for the electric-current distribution thereof  
25 from its result, the test apparatus 1300 has the antenna control circuit 1205 that includes a switch used as follows: In order to perform the rough irradiation at first, and then in order to make the



detailed test after identifying the region of location  
in question, the switch is used at first for cutting  
off the directionality-adjusting devices 102 of the  
narrow-directivity probe 1203 to transform the narrow-  
5 directivity probe into the common probe, and, at the  
time of the detailed test, the switch is used for  
connecting the directionality-adjusting devices 102  
thereto to transform the common probe back to the  
narrow-directivity probe. This antenna control circuit  
10 1205 is controlled using the computer 1211 or the like.  
Here, the operation state of the to-be-tested target  
1202 such as the electronic appliance at the time of  
irradiating the to-be-tested target with the  
electromagnetic field is inspected by a tester or a  
15 measurement device 1302 controlled using the computer  
1211 or the like. Moreover, its result is inputted  
into the computer 1211 or the like so as to make the  
test judgment.

In the apparatus for measuring the electric-  
20 field and/or magnetic-field distribution generated by  
an electronic appliance or the like, and for searching  
for the electric-current distribution of the electronic  
appliance or the like from its result, or in the test  
apparatus or the like for irradiating an electronic  
25 appliance or the like with an electric field and/or a  
magnetic field, and for observing the reaction from the  
electronic appliance or the like caused by this  
irradiation, there is provided a probe whose

directionality is narrower as compared with the directionality of the conventional probe. This makes it possible to provide the measurement/test apparatus exhibiting a tremendously high position resolution.

5           It should be further understood by those skilled in the art that although the foregoing description has been made on embodiments of the invention, the invention is not limited thereto and various changes and modifications may be made without  
10 departing from the spirit of the invention and the scope of the appended claims.